

MR4 RECEIVER

CIRCUIT DESCRIPTION

The receiver RF input circuit consists of seven high-Q helical resonators (H1 - H7), two amplifier transistors (Q1, Q2) and associated components. The gain of this circuitry at the RF input frequency is nominally 20 dB. Superior intermodulation performance is obtained by passing input signals through two helical resonators (H1, H2) to reject out-of-band energy before amplification. Both amplifier transistors use high bias current for maximum overload capability and minimum distortion. Feedback is used to stabilize amplifier operation against temperature variations, and two sections of power supply decoupling per stage, further insure stable operation. The RF circuit output connects to the input of double balanced mixer SBL-1.

Mixer injection voltage is generated by oscillator and multiplier stages consisting of transistors Q3 through Q6 and associated components. Q3 functions as a fundamental frequency oscillator at a frequency determined by crystals Y1 through Y4. In single frequency receivers, diode CR2 is replaced by a strap to cause crystal Y2 to be selected as the frequency determining element. Y2 may be enclosed by an optional proportional crystal oven in applications at UHF frequencies where the receiver is subject to wide temperature variations. In multi-frequency receivers, oscillator frequency is determined by providing a ground on terminal E5, E8, E11, or E14 to select the associated crystal. The multiplication of ratios of Q4 through Q6 depend upon the frequency range of the receiver as follows:

<u>Frequency Range</u> <u>in MHz</u>	<u>Q4</u>	<u>Multiplier</u> <u>Q5</u>	<u>Q6</u>
136 - 151	Doubler	Amplifier	Not Used
151 - 174	Tripler	Amplifier	Not Used
216 - 250	Doubler	Doubler	Not Used
420 - 512	Doubler	Doubler	Doubler

High frequency crystal are used in the MR4 receiver to minimize the number of possible image frequencies by reducing the total multiplication ratio needed to obtain the required injection frequency. To further reduce image levels, double-tuned filters are used between all multiplier stages.

The double balanced mixer output is fed to the high-IF amplifier section which consists of two transistors (Q7, Q8), eight crystal filter sections, and associated components ("B" version command receivers use four filter sections). The high-IF operates at a frequency of 21.4 MHz. Both amplifier stage outputs include broadly tuned resonant circuits (L24, C60 and L27, C66 respectively) to reject signals at frequencies beyond the skirts of the ceramic filters. Generous feedback and decoupling desensitize the amplifiers to temperature and power supply effects.

Conversion from high to low IF frequencies, amplification at the low IF frequency, limiting, and detection is done by integrated circuit U1 (squelch circuitry in U1 is not used). Frequency conversion is controlled by crystal Y5 which operates in conjunction with oscillator circuitry contained in U1. Four pole ceramic filter FL-9 operates at the low IF frequency of 455 KHz to provide additional filtering. Seven amplifier stages contained in U1 provide excellent limiting before detection. Detection is done by discriminator circuitry contained in U1 operating in conjunction with coil L28.

Wideband demodulated audio from U1 is detected by diodes CR5 and CR6 to provide the primary voltage reference for squelch operation. This detected voltage is fed to a Schmitt trigger circuit consisting of transistors Q15 and Q16. Hysteresis in the Schmitt trigger produces positive squelch action by requiring a change of about 6 dB in noise level before receiver audio is switched from off to on. The detected control R84 (R84 is external to the MR4). Action of the Schmitt trigger can also be controlled by an external CTCSS decoder to disable receiver audio when no CTCSS signal is present. In applications where CTCSS operation is used, the CTCSS decoder output connects to terminal E33 to control Schmitt trigger operation.

Squelch operation is further enhanced in the MR4 by automatically adjusting the squelch threshold in accordance with received signal level. Received signal at the low IF frequency is amplified by linear amplifier U2, detected by CR13/CR14, level shifted by U3A, and linear fed to fast/slow squelch switch transistor Q14. When weak signals, less than 1 μ V, are applied to the receiver transistor Q14 is turned on and applies a ground at terminal E30 to produce normal squelch action. When strong signals, greater than 1 μ V, are applied to the receiver transistor, Q14 is switched off. With Q14 turned off, the reference voltage at the Schmitt trigger input is increased causing the squelch to be "tightened". Tightening the squelch causes faster operation in response to signal changes and virtually eliminates squelch tail noise. Thus, the MR4 provides high squelch sensitivity to weak signals and noise free operation for strong signals.

Output from the Schmitt trigger gates the audio output of U1. When a received signal is present, audio from U1 is passed to amplifier U3C. Line audio is taken from the output of U3C at terminal E22. Audio from U3C is also routed to power amplifier U4 for driving a local speaker.

Metering circuits are provided to monitor signal strength, discriminator centering, and received signal peak deviation (metering is not provided in version "B" command receivers). All metering circuits are designed to drive 0-1 mA, 2200 Ohm panel meters. Signal strength metering is available at terminal E27 which is driven by DC amplifier transistor Q13. The signal strength meter indication is calibrated using potentiometer R66. Amplifier transistor Q11 drives terminal E20 to provide discriminator metering. Potentiometer R54 allows the discriminator meter to be set to mid-scale when registering an on-frequency signal. The discriminator meter negative terminal should be returned to ground through two series of connected diodes. Diodes CR7 and CR8 are provided external to the receiver module for this purpose when the MR4 is factory installed in a repeater or rack panel. Audio is amplified, rectified by CR9/CR10, and level shifted by Q12 to drive peak deviation metering output terminal E6. Metering calibration is done using potentiometer R64.

Switched outputs indicating the presence of received signal are available from Q9 or Q10. Q10 provides a ground at terminal E16 when received signal is present and an open when no signal is present. An inverted output can be obtained by connecting a strap from terminal E16 to terminal E17 and taking the output from terminal E15.

INSTALLATION

The following describes connections which may be made to MR4 receivers furnished in modular form. Receivers furnished in repeaters are completely connected and require no field installation.

1. Connect to receiver terminals E34 (positive) and E35 (ground) from a DC power source having the following characteristics:

Normal Voltage	12 VDC
Regulation	+/- 5%
Ripple	< 100 mV
Current	250 mA

CAUTION: The MR4 uses negative ground and must be powered from a negative ground or floating power supply. DO NOT connect a positive ground power source to the receiver.

2. To obtain a fixed-level audio output (line audio) connect to terminals E22 (signal) and E23 (ground). This output provides a level of approximately 0.4 vrms and a source impedance of 1000 Ohms. Shielded wire should be used for making this connection.
3. To obtain power amplifier audio output for driving a loudspeaker, connect to terminals E24 (signal) and E25 (ground). This output provides a fixed level of approximately 3 vrms. An external 100 Ohm potentiometer may be connected between this output and the loudspeaker, as shown in the receiver schematic drawing, to adjust loudspeaker volume.
4. Connect a strap from terminal E28 to terminal E29. (Input terminal E28 is used for remote squelch control in repeater applications and is not normally used in other applications.
5. Connect a 50 Kohm potentiometer to terminals E30, E31 and E32, as shown in the receiver schematic diagram. This potentiometer is used for setting the receiver squelch threshold. Shielded wire should be used for making these connections.
6. To obtain a logic output signal indicating the presence of received signal connect to terminal E16. Output E16 provides a ground when received signal is present and an open when no received signal is present.

If an inverted logic signal is desired, connect a strap from terminal E16 to terminal E17 and take the output from terminal E15. Terminal E15 provides an open when received signal is present and a ground when no received signal is present.

7. If CTCSS operation is to be used, connect from terminals E18 (signal) and E19 (ground) to the CTCSS decoder input. Connect from the CTCSS decoder output to terminal E33. The logic signal connected to terminal E33 should provide a ground when a CTCSS signal is detected by the decoder and an open when no CTCSS signal is detected.
8. If metering is to be used, connect 0-1 mA meters having internal resistances of 2200 Ohms to the following terminals:

	Meter + Terminal	Meter - Terminal
Signal Strength	E27	Ground
Peak Deviation	E26	Ground
Discriminator	E20	Ground through series connected diodes

9. Adjust the signal strength meter by applying a strong (10,000 uV) signal to the receiver input and setting potentiometer R66 until compression begins (further rotation of the potentiometer causes no further change in the meter indication).

Set the discriminator meter by applying an on frequency signal to the receiver and adjusting potentiometer R54 for a center scale meter reading.

Set the peak deviation meter by applying a signal having 5KHz deviation to the receiver and adjusting potentiometer R64 for a center scale reading. Meter indication is directly proportional to deviation, i.e., 0.5 mA indication corresponds to 5 KHz deviation.

ALIGNMENT

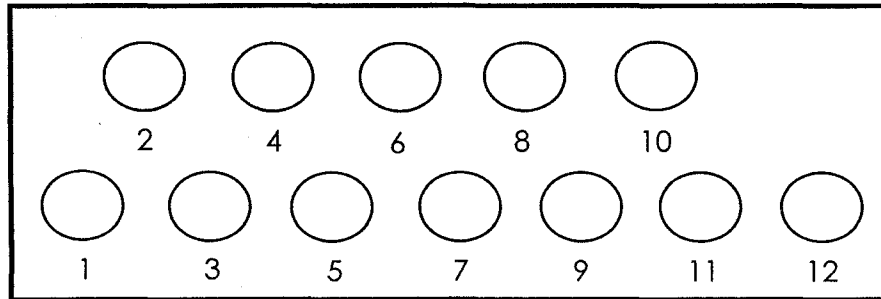
All MR4 receivers are factory aligned prior to shipment and require no initial alignment. The information given below is intended to aid re-adjustment following component replacement; it is not a step-by-step alignment procedure. Perform only the adjustments pertaining to the receiver section having the replaced component.

1. The RF section may be aligned by applying a signal to the receiver input and connecting a selective RF monitor to terminal E1. Adjust helical resonator capacitors and coupling capacitors C4 and C15 to obtain a maximum indication on the monitor. Some interaction will be found between resonator H2 and capacitor C4 and between resonator H6 and capacitor C15, so these adjustment should be repeated as necessary until no further increase in gain is obtained.

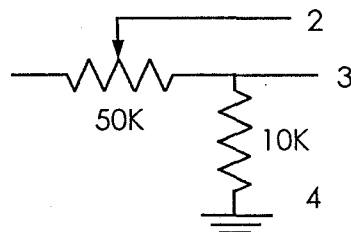
As adjustments are made, reduce the applied signal level to avoid saturating the amplifier transistors or the RF meter. When properly aligned, the RF section will exhibit a gain of approximately 20 dB.

2. If amplifier resonant circuits have a low Q, inductors L24 and L27 should not require adjustment in the field. The IF frequency can be set by loosely coupling a frequency counter to U1 terminal 1 and adjusting C107 to obtain a reading of 21.855 MHz.
3. The discriminator may be adjusted by applying an on-frequency, deviated signal to the receiver input and adjusting L28 for maximum output audio. The adjustment of L28 is not critical and will be found to have a broad maximum.
4. Receiver frequency may be set by applying an on-frequency signal to the receiver and adjusting capacitor C24 to obtain a center scale reading on the discriminator meter. In multi-frequency receivers, apply a ground to terminals E14, E11 E5, and E8 in turn to select the appropriate crystal and adjust capacitors C28, C26, C22 and C24, respectively.
5. Multiplier stages may be adjusted by connecting a selective RF monitor to test point TP4 and adjusting the tuning components (inductors and capacitors) for a maximum indication on the monitor. When the multipliers are operating properly, the injection voltage at TP4 should be greater than 300 mV.

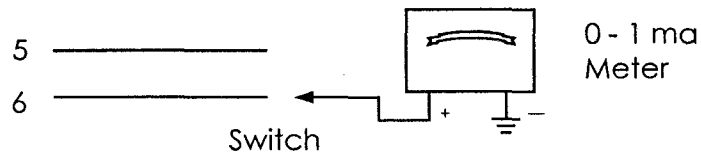
MR4 Receiver



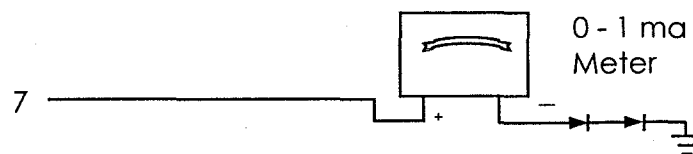
- 1. Squelch High
- 2. Squelch Wiper
- 3. Squelch Low
- 4. Ground



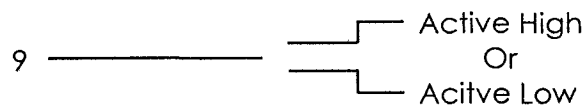
- 5. S Meter
- 6. Peak Deviation



- 7. Discriminator Meter
- 8. No Connection

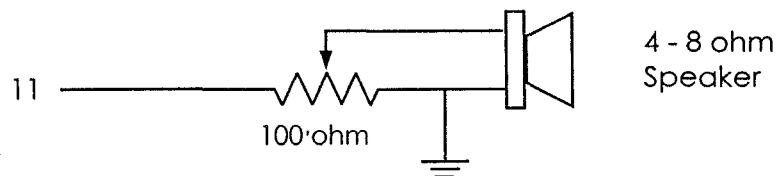


- 9. COS



- 10. Repeat Audio

- 11. Local Audio



MR4 RECEIVER PARTS LIST - 1 REV A

CAPACITORS

C 1	Part of Helical Resonator
C 2	Part of Helical Resonator
C 3	Part of Helical Resonator
C 4	5 - 25 pf Trim Cap
C 5	.001 Disc
C 6	.001 Disc
C 7	.001 Disc
C 8	.001 Disc
C 9	Part of Helical Resonator
C 10	Part of Helical Resonator
C 11	part of Helical Resonator
C 12	Part of Helical Resonator
C 13	Part of Helical Resonator
C 14	Part of Helical Resonator
C 15	5 - 25 pf Trim Cap
C 16	.001 Disc
C 17	.001 Disc
C 18	.001 Disc
C 19	.001 Disc
C 20	Part of Helical Resonator
C 21	.001 Disc
C 22	5 - 25 pf Trim Cap
C 23	.001 Disc
C 24	5 - 25 pf Trim Cap
C 25	.001 Disc
C 26	5 - 25 pf Trim Cap
C 27	.001 Disc
C 28	5 - 25 pf Trim Cap
C 29	.001 Disc
C 30	100 pf Silver Mica
C 31	4.7Mfd 126v.
C 32	.01 Disc
C 33	F.S. 8 pf SM (135 - 174 MHz) 8 pf SM (420 - 512 MHz) 12 pf SM (216 - 250 MHz)
C 34	F.S. 56 pf NPO (135 - 151 MHz) 47 pf NPO (151 - 158 MHz) 36 pf NPO (158 - 174 MHz) 39 pf NPO (215 - 250 MHz) 20 pf NPO (420 - 512 MHz)
C 35	.001 Disc
C 36	.001 Disc

RESISTORS

R 1	1K
R 2	1.8K
R 3	220 Ohm with F.Bead
R 4	100 Ohm
R 5	1K
R 6	1.8K
R 7	220 Ohms with F.Bead
R 8	100 Ohm
R 9	10K *
R 10	10K *
R 11	10K *
R 12	10K *
R 13	15K
R 14	3.3 K
R 15	220 Ohm
R 16	39 Ohm with F.Bead
R 17	1.5K
R 18	6.8K
R 19	1.8K
R 20	220 Ohm
R 21	47 Ohm with F.Bead
R 22	6.8K
R 23	1.8K
R 24	220 Ohm
R 25	47 Ohm with F.Bead
R 26	6.8K
R 27	1.8K
R 28	220 Ohm
R 29	47 Ohm with F.Bead
R 30	1K
R 31	2.7K
R 32	1K
R 33	47 Ohm with F.Bead
R 34	75 Ohm
R 35	2K
R 36	2K
R 37	47 Ohm with F.Bead
R 38	1K
R 39	2.7K
R 40	1 K
R 41	75 Ohm
R 42	2.2K
R 43	100K
R 44	4.7K

C 37 F.S. 1 pf tubular
CAPACITORS

C 38 F.S.
10 pf NPO (135 - 151 MHz)
12 pf NPO (151 - 158 MHz)
10 pf NPO (158 - 250 MHz)
15 pf NPO (420- 512 MHz)

C 39 F.S.
10 pf NPO (135 - 151 MHz)
10 pf NPO (151 - 158 MHz)
8 pf NPO (158 - 174 MHz)
12 pf NPO (215 - 512 MHz)

C 40 1 pf (5 pf 220 MHz only)

C 41 .001 Disc

C 42 .01 Disc

C 43 F.S. 1 pf tubular

C 44 F.S.
22 pf (135 - 151 MHz)
(151 - 158 MHz)
(158 - 174 MHz)
(215 - 250 MHz)
(420 - 512 MHz)

C 45 6.8 NPO Disc

C 46 F.S. 1 pf

C 47 .001 Disc

C 48 .01 Disc **

C 49 3 -15 pf Trim Cap **

C 50 1 pf tubular

C 51 3 -15 pf Trim Cap **

C 52 3 -15 pf Trim Cap **

C 53 3 -15 pf Trim Cap **

C 54 1 pf tubular **

C 55 33 pf NPO

C 56 10 pf NPO

C 57 27 pf NPO

C 58 .01 Disc

C 59 .01 Disc

C 60 33 pf NPO

C 61 .01 Disc

C 62 10 pf NPO

C 63 .01 Disc

C 64 .01 Disc

C 65 .001 Disc

C 66 33 pf NPO

C 67 .01 Disc

C 68 100 pf S.M.

C 69 17 pf S.M.

C 70 .01 Disc

C 71 100 pf NPO

C 72 .1 Disc

R 45 1K
RESISTORS

R 46 47K

R 47 4.7K

R 48 390K

R 49 10K

R 50 20K

R 51 7.5K

R 52 4.7K

R 53 390 Ohm

R 54 10K Pot

R 55 20K

R 56 47K

R 57 20K

R 58 1K

R 59 200K

R 60 39 K F.S.

R 61 470K

R 62 1 Meg

R 63 470K

R 65 1 Meg

R 66 5K Pot

R 67 22 meg

R 68 150 Ohm with F.Bead

R 69 22 Meg

R 70 3.9K

R 71 2.2K

R 72 470K

R 73 10K

R 74 1 Meg

R 75 10K

R 76 20K

R 77 4.7K

R 78 100K

R 79 10K

R 80 10K

R 81 10K

R 82 150 Ohm

R 83 3.9K

R 84 50K pot (Squelch)

SEMI-CONDUCTOR DEVICES

C 73 .1 Disc
C 74 100 pf S.M.
C 75 100 pf NPO
C 76 100 pf NPO
C 77 4.7 MFD 16 v.
C 78 4.7 MFD 16v.
C 79 .047 Mylar
C 80 .01 Mylar
C 81 2.2 MFD
C 82 2.2 MFD 16 v.
C 83 4.7 MFD 16 v.
C 84 220 MFD 35 v.
C 85 .
C 86 33 MFD 35 v.
C 87 .1 MFD Disc
C 88 220 MFD 35 v.
C 89 .1 MFD Disc
C 90 .1 MFD Disc
C 91 .1 MFD Disc
C 92 100 pf NPO
C 93 100 pf NPO
C 94 4.7 MFD 16v.
C 95 100 pf NPO
C 96 100 pf NPO
C 97 2.2 MFD 16v.
C 98 .1 MFD Disc
C 99 .1 MFD Disc
C 100 .1 MFD Disc
C 101 .1 MFD Disc
C 102 2.2 MFD 16 v.
C 103 33 MFD 35 v.
C 104 .1 MFD Disc
C 105 .01 Mylar
C 106
C 107 5 - 25 pf Trim Cap

CRYSTALS AND FILTERS

Y1 - Y4

136 - 151 Fx = (FO-21.4Mhz)/2
151 - 174 Fx = (FO-21.4Mhz)/2
216 - 250 Fx = (FO-21.4Mhz)/4
420 - 512 Fx = (FO-21.4MHz)/8
Parallel resonant, third overtone,
resistance 30 Ohm max. 12 pf load
capacity, HC-25/U case

Q 1 MRF 901
Q 2 MRF 901
Q 3 2N4123
Q 4 2N3563
Q 5 2N3563
Q 6 2N5179
Q 7 2N3563
Q 8 2N3563
Q 9 2N4123
Q 10 2N4123
Q 11 2N4123
Q 12 2N4123
Q 13 2N4123
Q 14 2N4123
Q 15 2N4123
Q 16 2N4123

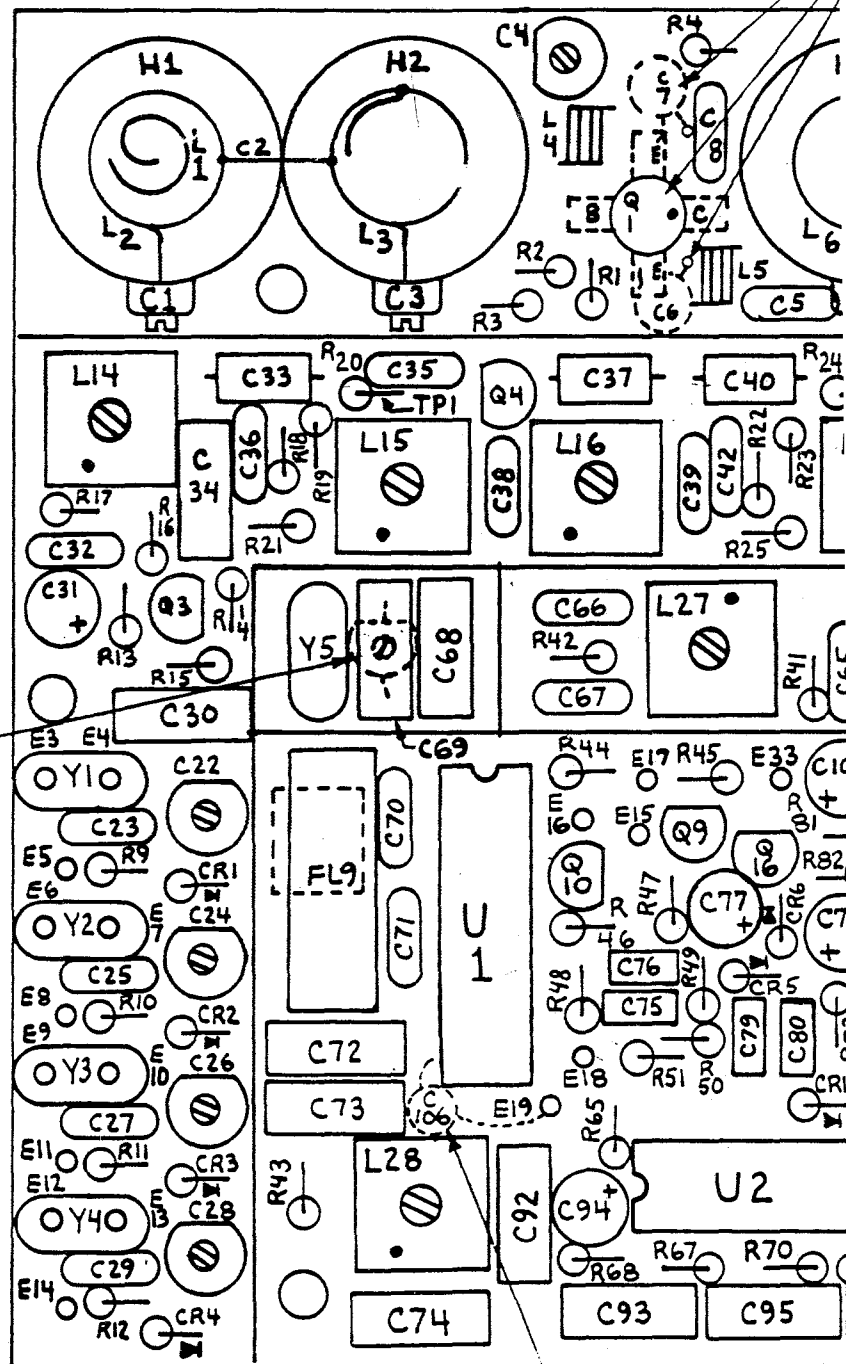
U 1 MC 3359
U 2 CD4007
U 3 LM3900
U 4 LM380
U 5 7808

CR 1 1N1448 *
CR 2 1N1448 *
CR 3 1N1448 *
CR 4 1N1448 *
CR 5-14 1N1448

COILS AND INDUCTORS

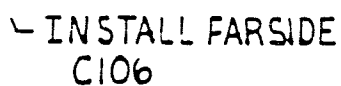
L1 Part of Helical Resonator
L2 Part of Helical Resonator
L3 Part of Helical Resonator
L4 6T2518 (135 - 174 MHz)
L5 Part of Helical Resonator
L6 Part of Helical Resonator
L7 Part of Helical Resonator
L8 Part of Helical Resonator
L9 Part of Helical Resonator
L10 6T2528 (135 - 174 MHz)
5T2518 (216 - 250 MHz)
L11 5T2518
L12 Part of Helical Resonator
L13 Part of Helical Resonator

INSTALL FAR SIDE
C107



INSTALL FAR SIDE
C106

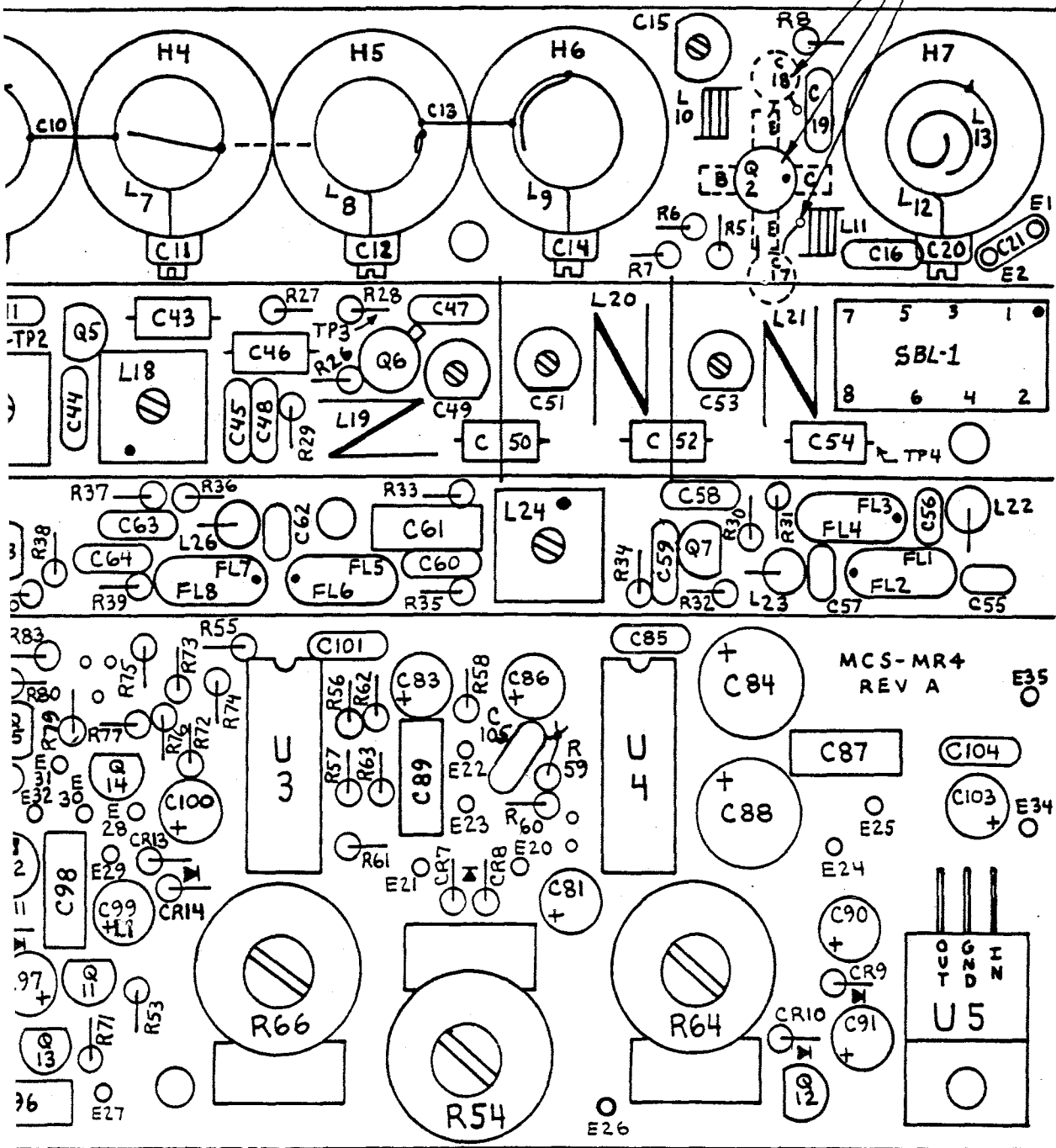
Q1, C6, C7



✓ INSTALL FAR SIDE
C106

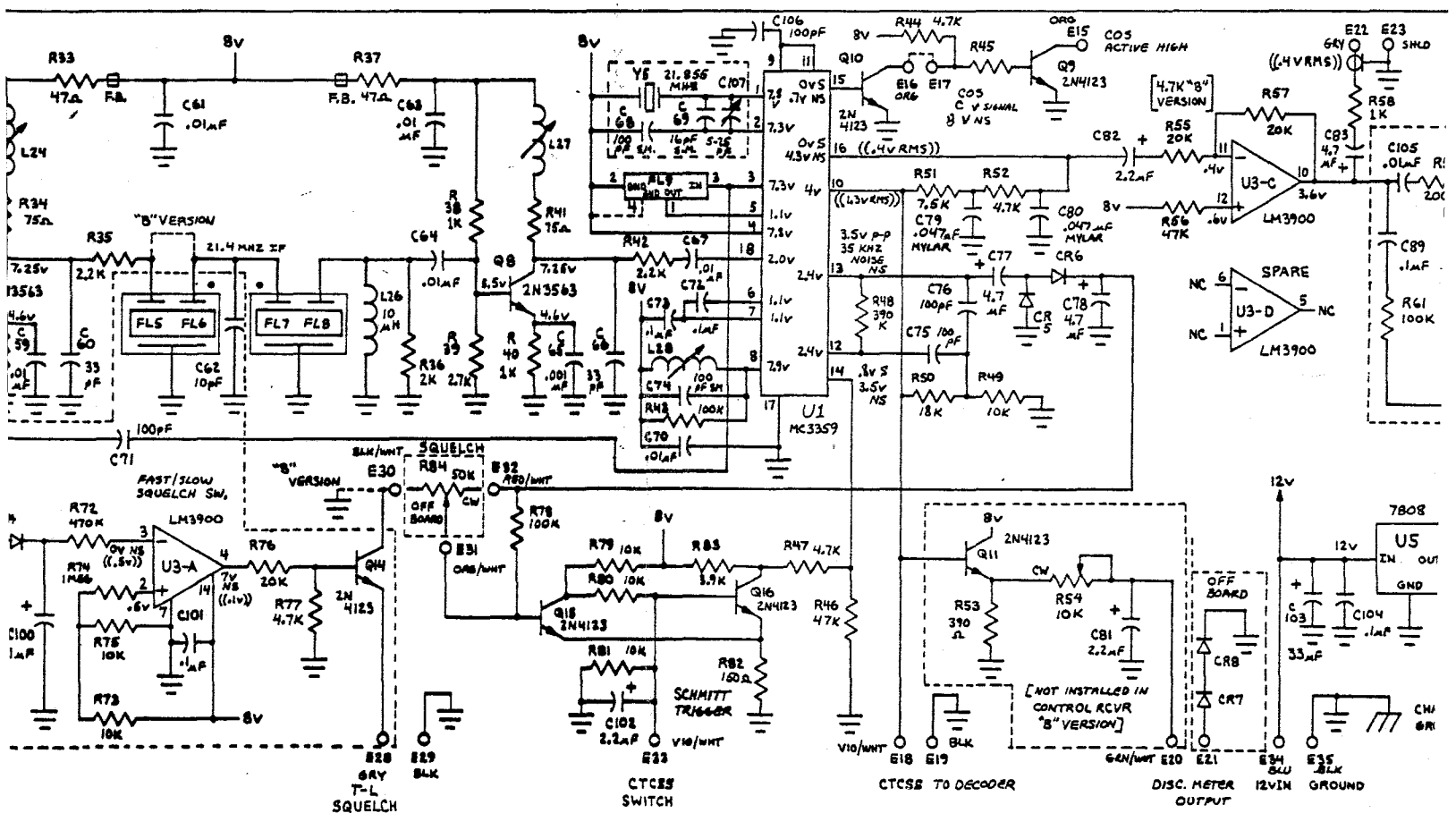
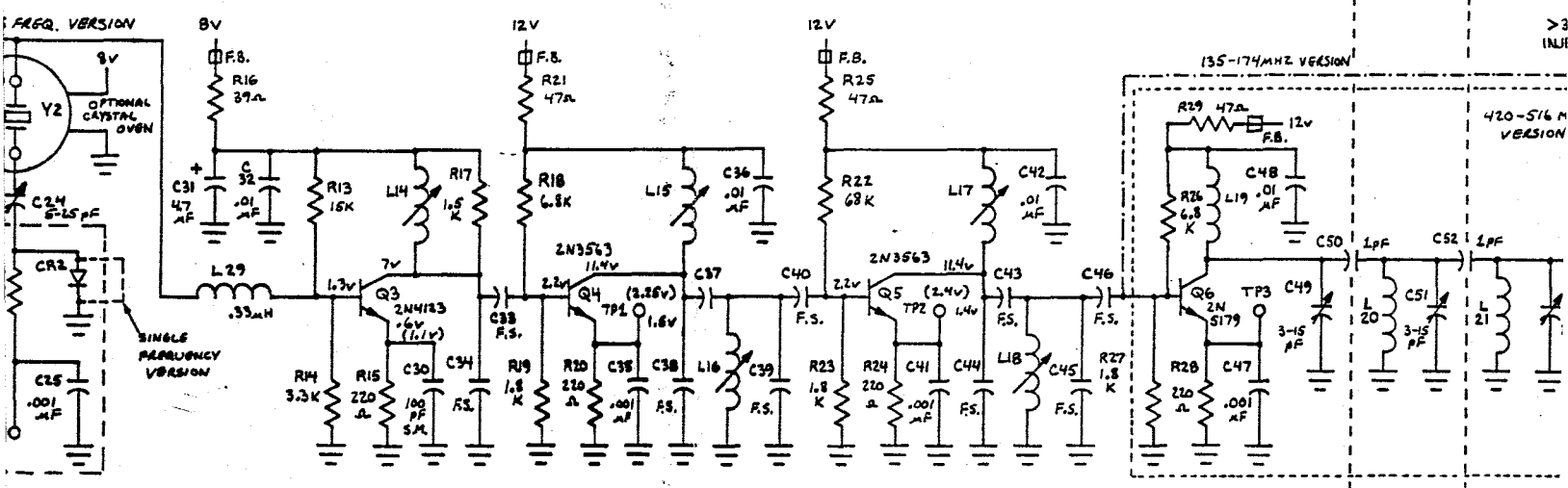
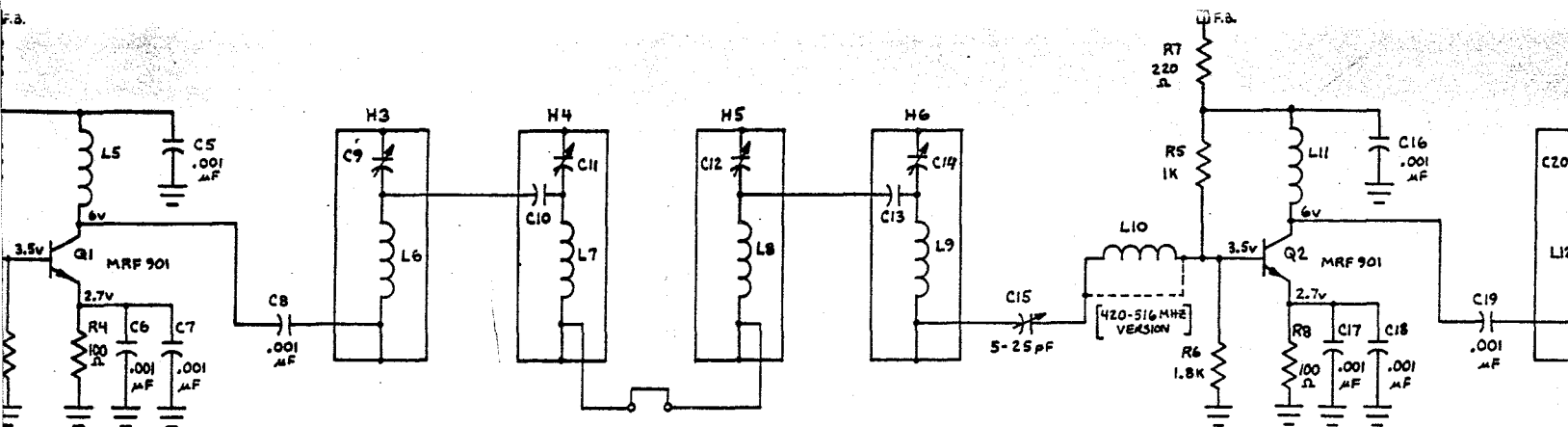
INSTALL FAR SIDE
Q1, C6, C7

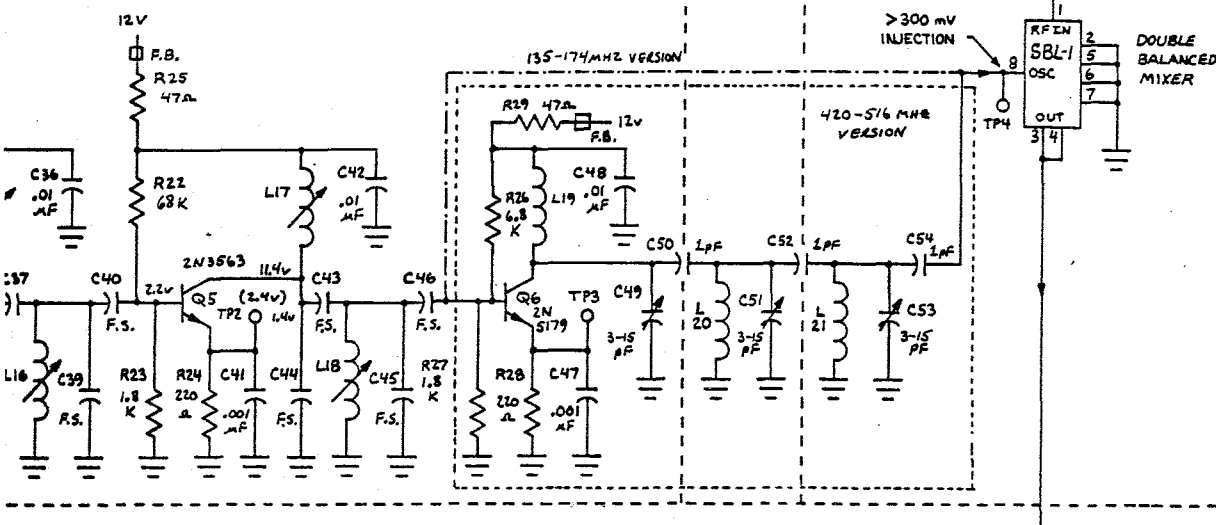
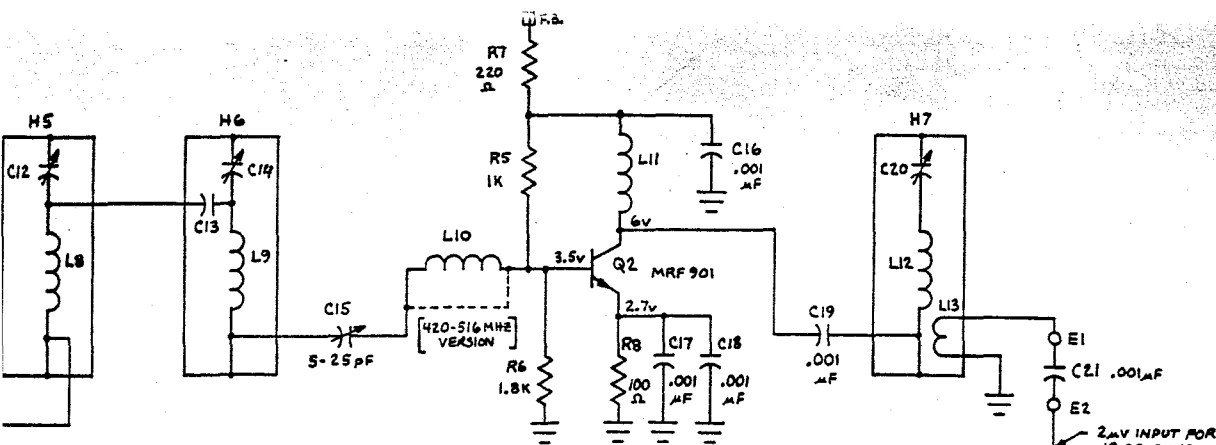
INSTALL FAR SIDE
Q2, C17, C18



MICRO CONTROL SPECIALTIES	
23 ELM PARK GROVELAND, MA. 01834	
2 x 1	8 23 83 WILLIAM R. DEWHIRST DRAGONWYCK
ASSEMBLY DRAWING	
RECEIVER P.C. BOARD	REV A3 MCS - MRR

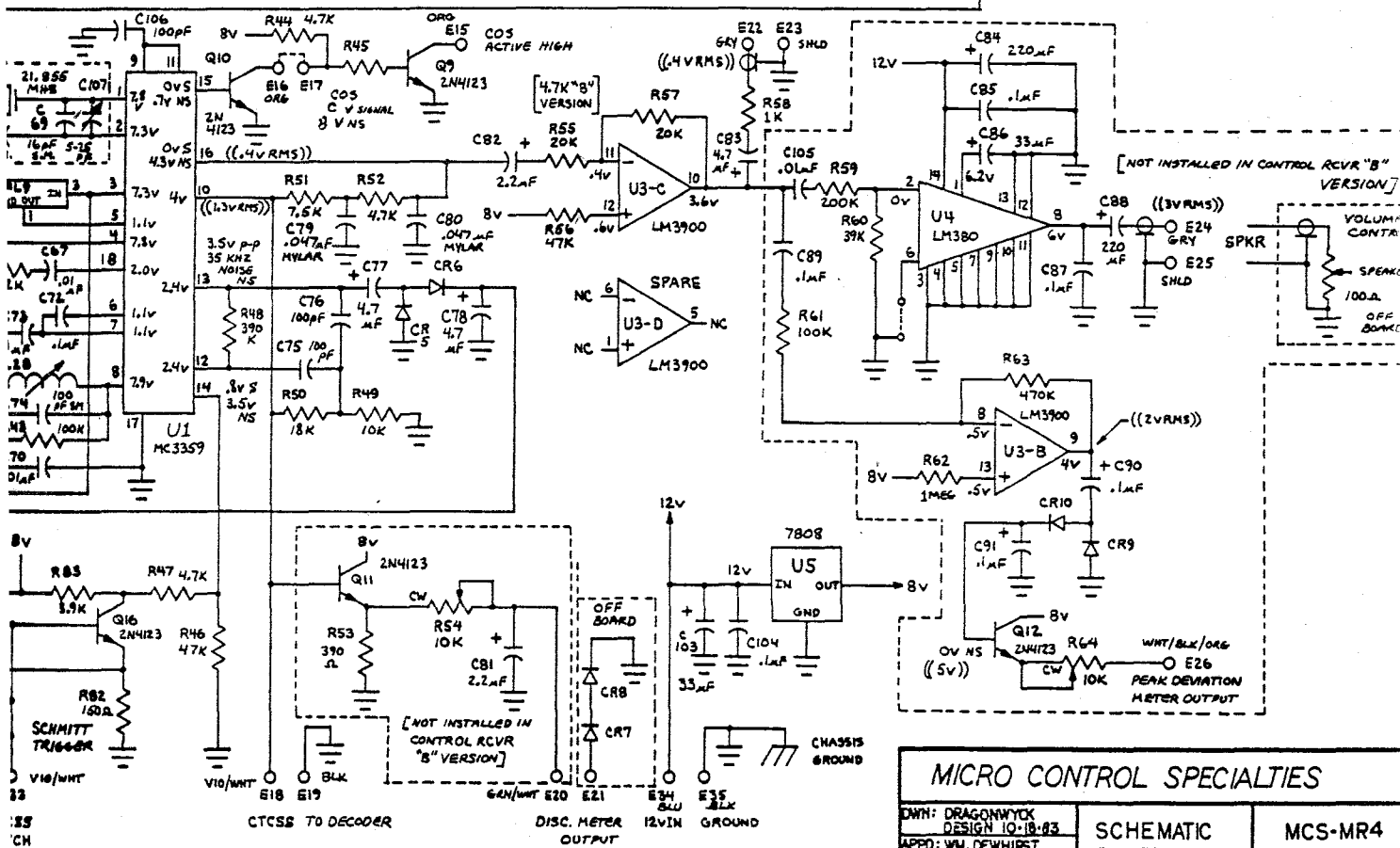
F.A.





NOTES:

1. VOLTAGE MEASUREMENTS T ON A 18 MEG. INPUT DVM.
2. VOLTAGES MEASURED WITH S = RF SIGNAL INPUT. NS = NO RF SIGNAL INPUT. () = CRYSTAL INSTALLED. () = 100mV RF INPUT WITH A 1 KHZ SIGNAL DEVIATED ± 5 KHZ.
3. UNLESS OTHERWISE SPECIFIED, ALL DIODES ARE 1N4148.
4. H1-H7 ARE HELICAL RESONATORS.



MICRO CONTROL SPECIALTIES

OWN: DRAGONWYCK
DESIGN 10-18-83
APPD: WM. DEWHIRST

SCHEMATIC
RECEIVER

MCS-MR4